

nlnmul.exe Exercises for Chapter nlnmul

Esercizio nlnmul010000

Write a one- or two-sentence response to each of the following questions and imperatives. The use of equations is acceptable when they appear in a sentence. Don't quote me (use your own words, other than technical terminology).

- (a) Write the equivalent impedance of a resistor R and an inductor L in series. Express the result in rectangular and polar (phasor) form.
- (b) How do you find the Norton equivalent resistance?
- (c) Explain how a diode operates in forward-bias.
- (d) In a MOSFET, how much current will flow from the drain D to the source S when the gate-source voltage is 0.3 V ? Succinctly explain/justify.

Esercizio nlnmul010010

Write a one- or two-sentence response to each of the following questions and imperatives. The use of equations is acceptable when they appear in a sentence. Don't quote me (use your own words, other than technical terminology).

- (a) Describe a couple differences between MOSFETs and opamps.
- (b) If a DC source is connected to a circuit in steady state, describe how an inductor in the circuit will be operating.
- (c) If a transformer increases an AC signal's voltage by a factor of 10, what happens to the signal's current?
- (d) How do we determine the diode resistance for the piecewise linear model of a diode?

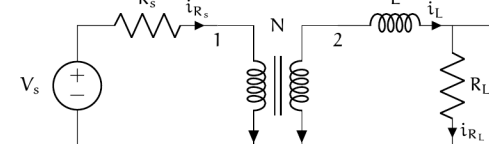


Figure exe.1: circuit diagram for Exercise nlnmul010010 and Exercise nlnmul010020.

Esercizio nlnmul010020

Write a one- or two-sentence response to each of the following questions and imperatives. The use of equations is acceptable when they appear in a sentence. Don't quote me (use your own words, other than technical terminology).

- (a) If the current through an inductor is suddenly switched off, what happens?
- (b) Let the output voltage of a resistor circuit be 5 V and the equivalent resistance 50Ω . What is the Thevenin equivalent circuit?
- (c) In the preceding part of this question, what is the Norton equivalent?
- (d) When can we use impedance analysis?

Esercizio nlnmul010030

For the circuit diagram of Fig. exe.1, solve for $v_o(t)$ if $V_s(t) = A \cos \omega t$. Let $N = n_2/n_1$, where n_1 and n_2 are the number of turns in each coil, 1 and 2, respectively. Also let $i_o(t) = i$ be the initial condition.

Esercizio nlnmul010040

Reads Exercise nlnmul010030, but only consider the steady-state response. Use impedance methods!

Esercizio nlnmul010050

Calculate the current through a diode using the ideal model under the following conditions.

$v_D = 5.8 \text{ V}$
 $T = 38.21, 28^\circ\text{C}$

The diode can be assumed to have a saturation current of $i_s = 10^{-12} \text{ A}$. You may find the following helpful.

- Boltzmann constant: $1.38 \times 10^{-23} \frac{\text{J}}{\text{K}}$
- fundamental charge: $1.602 \times 10^{-19} \text{ C}$.

Esercizio nlnmul010060

When considering the steady state of circuits with only DC sources, all voltages and currents are constant and all diodes are in constant states (each is ON or OFF). The methods of Lec. nlnmul0100 still apply, of course, but we needn't be concerned with a time evolution. Consider the circuits of Fig. exe.2. For each circuit, solve for the voltage across the $5 \text{ k}\Omega$ resistor. Treat each diode as an ideal diode.

Esercizio nlnmul010070

Repeat Exercise nlnmul010060, but use the piecewise linear model of each diode.

Esercizio nlnmul010080

A diode clipping circuit is one that "clips" the tops and/or bottoms of a signal. These circuits can be used to set a maximum or minimum voltage for a signal. Consider the diode clipping circuit of Fig. exe.3. Source V_1 effectively adjusts the maximum possible load voltage v_{oL} , and V_2 the minimum. Let $V_2(t) = 10 \cos \omega t$, $V_1 = 5 \text{ V}$, $V_2 = -3 \text{ V}$, and $R_L = R_1 = 50 \Omega$. Solve for $v_{oL}(t)$. Use the ideal diode model.

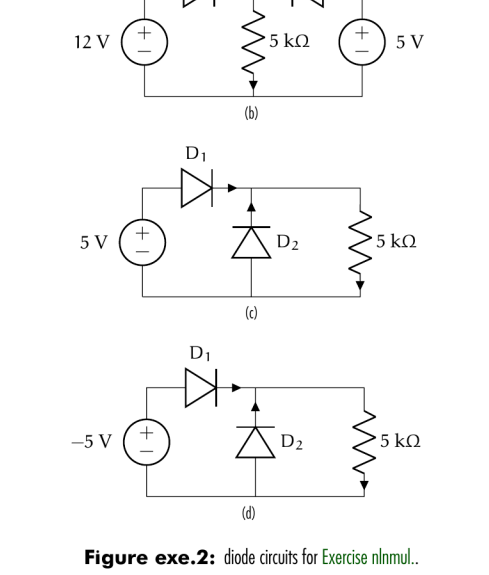


Figure exe.2: diode circuit for Exercise nlnmul010080.

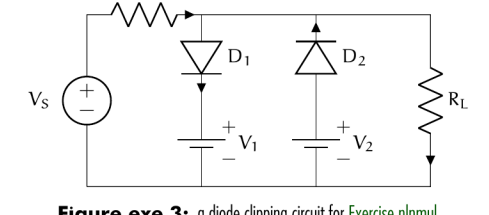


Figure exe.3: diode clipping circuit for Exercise nlnmul010080.

Esercizio nlnmul010090

Repeat Exercise nlnmul010080, but use the piecewise linear model of each diode.

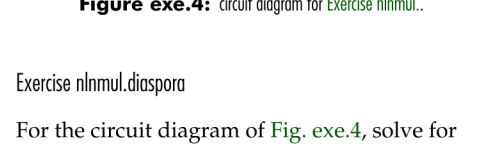


Figure exe.4: circuit diagram for Exercise nlnmul010090.

Esercizio nlnmul010100

For the circuit diagram of Fig. exe.4, solve for $v_o(t)$ if $V_s(t) = A$ for some given $A > 0.5 \text{ V}$. Let $v_o(t)|_{t=0} = 0 \text{ V}$ be the initial condition. Use a piecewise linear model for the diode with some $R_d \in \mathbb{R}_{>0}$. Do not estimate R_d .

Esercizio nlnmul010110

For the circuit shown in Fig. exe.5, determine the voltage across the load v_{oL} in terms of parameters and the gate voltage source voltage V_g and V_c . The parameters of the MOSFET are k_n and $V_{gs,th}$. Assume MOSFET saturation operation.

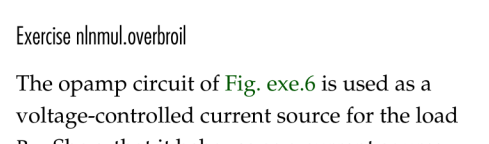


Figure exe.5: circuit for Exercise nlnmul010110.

Esercizio nlnmul010120

The opamp circuit of Fig. exe.6 is used as a voltage-controlled current source for the load R_L . Show that it behaves as a current source with current i_{oL} controlled by voltage source v_i . Use two separate methods: (a) assuming $v_i = v_i$, and (b) not assuming $v_i = v_i$, rather, assuming the open loop gain of the opamp A is large. Comment on the differences between the methods of (a) and (b).

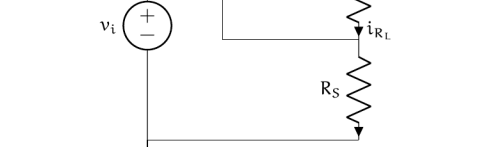


Figure exe.6: circuit for Exercise nlnmul010120.

Esercizio nlnmul010130

Use the circuit diagram of Fig. exe.7 to answer the questions below. Use the sign convention from the diagram. Let $v_i = A \cos \omega t$ be an ac input voltage. The load Z_L impedance is not given.

- (a) Write the elemental equations in terms of Z_L , Z_{R_1} , Z_{R_2} , and Z_L (the impedances of the components).
- (b) Write the KCL and KVL equations.
- (c) Solve for the steady-state $v_o(t)$ without inserting the values of the impedances (that is, leave it in terms of Z_L , Z_{R_1} , Z_{R_2} , and Z_L).

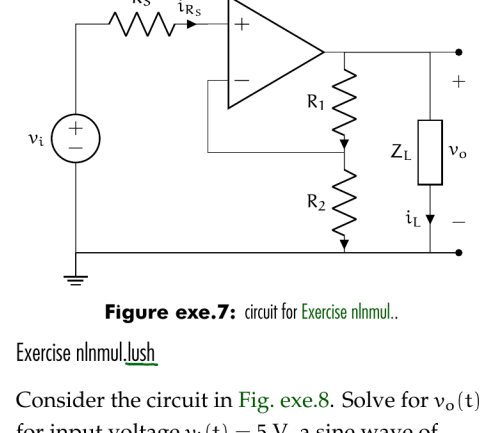


Figure exe.7: circuit for Exercise nlnmul010130.

Esercizio nlnmul010140

Consider the circuit in Fig. exe.8. Solve for $v_o(t)$ for input voltage $v_i(t) = 5 \text{ V}$ sine wave of $v_i(t) = 5 \sin 2525t$, and a sine wave of $v_i(t) = 5 \sin 2525t$. Let $R_1 = 50 \Omega$, $R_2 = 10 \text{ k}\Omega$, $C = 1 \mu\text{F}$, and the opamp open-loop gain be $A = 10^5$. Let the initial condition be $v_c(t) = 0 \text{ V}$. In each case, plot the solution to show the transient response until it reaches steady-state.

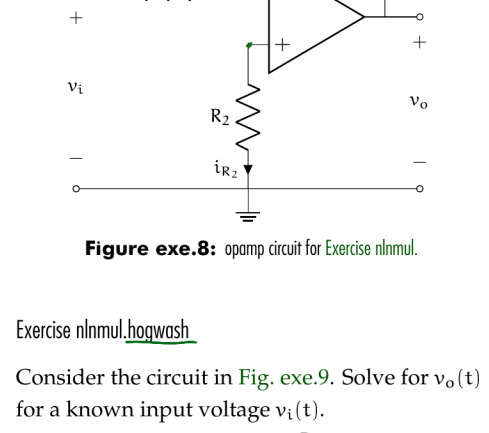


Figure exe.8: opamp circuit for Exercise nlnmul010140.

Esercizio nlnmul010150

Consider the circuit in Fig. exe.9. Solve for $v_o(t)$ for a known input voltage $v_i(t)$.

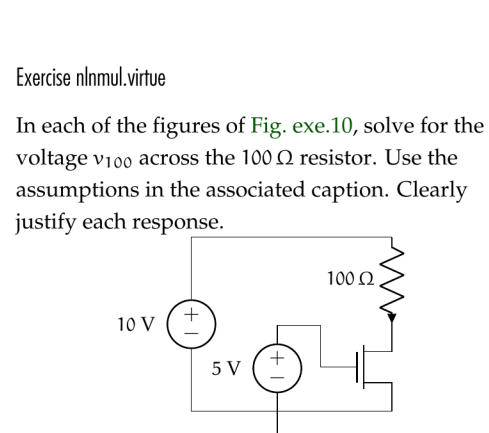


Figure exe.9: opamp circuit for Exercise nlnmul010150.

Esercizio nlnmul010160

In each of the figures of Fig. exe.10, solve for the voltage $v_{D,ss}$ across the 100Ω resistor. Clearly justify each response.

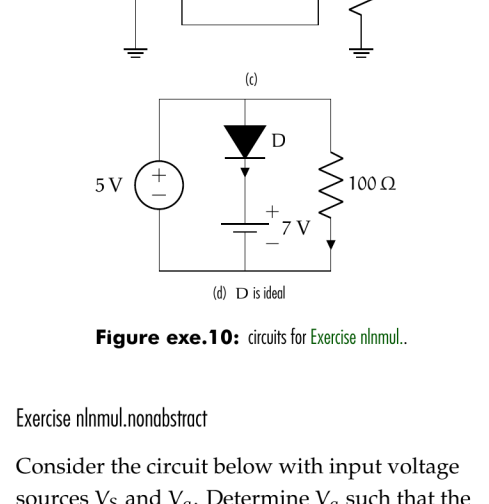
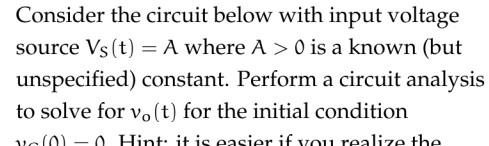


Figure exe.10: circuit for Exercise nlnmul010160.

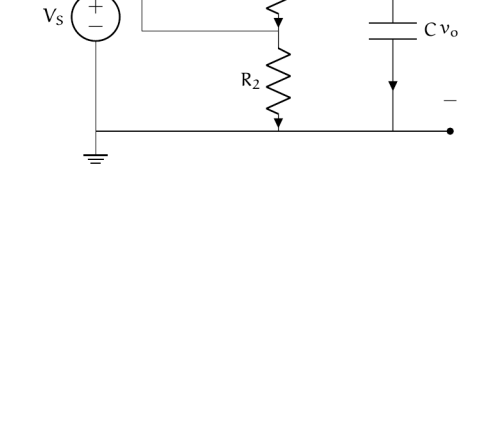
Esercizio nlnmul010170

Consider the circuit below with input voltage sources V_1 and V_2 . Determine V_o such that the load voltage $v_{oL} = 10 \text{ V}$. Let $R_1 = 2 \text{ k}\Omega$, $R_2 = 0.5 \text{ mA/V}^2$, $V_1 = 0.7 \text{ V}$, $V_2 = 20 \text{ V}$.



Esercizio nlnmul010180

Consider the circuit below with input voltage source $V_1(t) = A \cos \omega t$ where $A > 0$ is a known (but unspecified) constant. Perform a circuit analysis to solve for $v_o(t)$ for the initial condition $v_o(t) = 0$. Hint: It is easier if you realize the opamp output voltage is effectively an ideal voltage source (so it does not depend on v_o and v_c) and you can therefore treat the two parts of the circuit separately.



$$i_{oL} = v_{oL} / R_L$$

$$V_{oL} = i_{oL} R_L$$

$$= \left(\frac{V_{gs} - V_{gs,th}}{R_s} - \frac{k_n}{2} (V_{gs} - V_{gs,th})^2 \right) R_L$$

$$= \left(\frac{V_g - V_{oL}}{R_s} - \frac{k_n}{2} (V_g - V_{oL})^2 \right) R_L$$

$$V_{oL} + \frac{k_n R_L}{2} V_{oL}^2 = \left(\frac{V_g}{R_s} - \frac{k_n}{2} (V_g - V_{oL})^2 \right) R_L$$

$$V_{oL} \left(1 + \frac{k_n R_L}{2} \right) = \frac{R_L}{R_s} \left(\frac{V_g}{R_s} - \frac{k_n}{2} (V_g - V_{oL})^2 \right)$$

$$V_{oL} = \frac{R_L}{R_s + \frac{k_n R_L^2}{2}} \left(\frac{V_g}{R_s} - \frac{k_n}{2} (V_g - V_{oL})^2 \right)$$

$$= \frac{R_L}{R_s + \frac{k_n R_L^2}{2}} \left(\frac{V_g}{R_s} - \frac{k_n}{2} V_g^2 + \frac{k_n}{2} V_g V_{oL} - \frac{k_n}{2} V_{oL}^2 \right)$$

$$= \frac{R_L}{R_s + \frac{k_n R_L^2}{2}} \left(V_g - \frac{k_n R_s}{2} (V_g - V_{oL})^2 \right)$$

$$E(t) = i_{oL} R_L$$

$$= \frac{d v_{oL}}{dt} = \frac{1}{C} i_C$$

$$V_{oL} = i_C R$$

$$i_C = \frac{1}{R} V_{oL}$$

$$V_o = V_c - V_{oL} - V_{oL}$$

$$E(t) = i_{oL} R_L$$

$$= \frac{d v_{oL}}{dt} = \frac{1}{C} i_C$$

$$V_{oL} = i_C R$$

$$i_C = \frac{1}{R} V_{oL}$$

$$V_o = V_c - V_{oL} - V_{oL}$$

$$E(t) = i_{oL} R_L$$

$$= \frac{d v_{oL}}{dt} = \frac{1}{C} i_C$$

$$V_{oL} = i_C R$$

$$i_C = \frac{1}{R} V_{oL}$$

$$V_o = V_c - V_{oL} - V_{oL}$$